Technical insights into advanced numerical methods applied in 3D EM modelling

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MT & CSEM methods in frequency domain



Receiver station setup (Bücker et al., 2017) for land-based MT and CSEM measurements.

Magnetotellurics (MT) & Audiomagnetotellurics (AMT)

Natural sources f: $\approx 10^{-4} - 10^4$ Hz Far-field assumption (plane waves) Modelling: Source is impressed at domain boundaries or at anomalies

Controlled-source electromagnetics (CSEM)

Artificial sources f: $\approx 10^{-1} - 10^4$ Hz Modelling: Source has to be included in the modelling domain

Model parameters: ρ , μ , ϵ



Boundary value problem (BVP)

Total field formulation in terms of electric field **E** in frequency domain with time dependence $e^{i\omega t}$ for • **MT**:

$$\nabla \times \frac{1}{\mu} \nabla \times \mathbf{E} + i\omega\sigma\mathbf{E} - \omega^{2}\varepsilon\mathbf{E} = \mathbf{0} \quad \text{in} \quad \Omega,$$

$$\hat{\mathbf{n}} \times \frac{1}{-i\omega\mu} \nabla \times \mathbf{E} = \mathbf{g}_{t} \quad \text{on} \quad \partial\Omega,$$
 (1)

where ω is the angular frequency and $\mathbf{g}_t = \hat{\mathbf{n}} \times \mathbf{H}_0$, \mathbf{H}_0 could be the plane wave magnetic field solution for a 1D or 2D background model.

CSEM:

$$\nabla \times \frac{1}{\mu} \nabla \times \mathbf{E} + i \omega \sigma \mathbf{E} - \omega^2 \varepsilon \mathbf{E} = -i \omega \mathbf{J}_{source} \quad \text{in} \quad \Omega,$$
$$\hat{\mathbf{n}} \times \mathbf{E} = 0 \quad on \quad \partial \Omega,$$





(2)

Forward modelling

Compute electric and magnetic fields at receiver locations for a known model

Workflow:

- Formulate BVP
- Oecide on a discretisation approach for solving the BVP numerically
- Discretise the model domain
- Assemble the system of equations (SOE)
- Apply boundary conditions
- Solve the SOE for E or H on mesh nodes or edges
- Calculate fields at receiver stations





Motivation

Advancing standard approaches for discretisation, boundary methods and solving the SOE

Discretisation & boundary methods:

- Obtain better accuracy of forward responses
- Reduce problem sizes

Solve:

 Reduce computational cost (time and memory) per forward solution while assuring accuracy



Error sources associated with the numerical solution of a field problem (Mattiussi, 2001).



Motivation

Advancing standard approaches for discretisation, boundary methods and solving the SOE

Discretisation & boundary methods:

- Obtain better accuracy of forward responses
- Reduce problem sizes

Solve:

 Reduce computational cost (time and memory) per forward solution

... to speed up the inversion process



Topics



- Einite elements with mesh refinement
- Constrained allowants with high order internals
- Spectral elements with high-order interpolation
- Boundary Methods
 - Standard method: Dirichlet
 - Perfectly Matched Layers
- Solving
 - Direct Solvers
 - Iterative Solvers



Topics

Discretisation

- Finite elements with mesh refinement
- Spectral elements with high-order interpolation
- 2 Boundary Methods
 - Standard method: Dirichlet
 - Perfectly Matched Layers

Solving

- Direct Solvers
- Iterative Solvers



Discretisation

Numerical methods for discrete approximations:

- Finite difference method (FDM)
- Method of characteristics (MOC)
- Finite element method (FEM)
- Finite volume method (FVM)
- Boundary element method (BEM)
- Meshless method (MLM)
- Spectral element method (SEM)

Commonly used element shapes \rightarrow (Diersch, 2014)



Finite elements (FE)



Interpolation functions: 1st order **Nédélec** (Jin, 2014)

$$\mathbf{N}_i = (L_{i1}\nabla L_{i2} - L_{i2}\nabla L_{i1})I_i$$

Code: elfe3D (Rulff et al., 2021)

(modelling with the total **e**lectric field approach using **f**inite **e**lements in **3D**)

- Variable σ , μ , ϵ distribution
- Customisable mesh refinement: goal-oriented adaptive, uniform global mesh quality, combination of error estimators
- Direct solver: PARDISO (Schenk and Gärtner, 2004) or MUMPS (Amestoy et al., 2001)



Tetrahedral FE

Elements with the most inaccurate solutions are identified by high element error estimators: bisected, if they exceed a certain user-defined threshold.

Goal-oriented adaptive refinement (h-refinement):

- Based on error estimators formulated for a primal and an adjoint problem (Ren et al., 2013)
- 3 optional error estimator components:
 - Residual
 - Face jumps in normal current density
 - Face jumps in tangential magnetic field

$$\begin{split} \eta^{e}_{E} &= \sqrt{(r^{e}_{E})^{2} + (jf^{e}_{E})^{2} + (hf^{e}_{E})^{2}} \qquad (primal) \\ \eta^{e}_{W} &= \sqrt{(r^{e}_{W})^{2} + (jf^{e}_{W})^{2} + (hf^{e}_{W})^{2}} \qquad (adjoint) \end{split}$$



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Two elements, labelled 1 and 2, sharing one face. The continuity of (a) the normal component of the current density $J^{\vec{n}}$ and (b) the tangential component of the magnetic field $\mathbf{H}^{\vec{n}\times}$ cannot be guaranteed across the common interface using an edge-based finite-element formulation for the electric field. Hence, face jumps in $J^{\vec{n}}$ and $\mathbf{H}^{\vec{n}\times}$ are used as optional terms in the error estimator guiding the mesh refinement (Rulff et al., 2021).



Special elfe3D features for mesh refinement:

- Face jumps in tangential magnetic field can be included in the error estimator: helpful for models with magnetic permeability contrasts
- Refinement on low quality mesh option, increasing quality only in the last refinement step: saves time and memory
- Using average relative global error estimator η_G and average relative error estimator at the receivers η_B to assess refinement behaviour

$$\eta_G = \frac{1}{M} \sum_{e=1}^M \eta_L^e, \qquad (4)$$
$$\eta_R = \frac{1}{N_r} \sum_{r=1}^{N_r} \eta_L^r, \qquad (5)$$



Special elfe3D features for mesh refinement:

- Source strengths in the adjoint problem take the distance to the primal source into account: ensures that all receiver locations are included equally in the error estimation
- Element error estimators are weighted with the local field strengths: avoids amplitude-dependent over-refining of the mesh

weighting factors	decrease in η _G	decrease in η _R	receiver refinement	anomaly refinement	accuracy improvement
<i>r</i> ³ & amplitudes	1	1	1	1	1
r ³	×	×	×	×	×
amplitudes	1	×	×	\checkmark	×
none	×	×	×	×	✓(minor)



Example I FE & refinement for a synthetic block model



Vertical cuts of test models at y = 0 km. Left: homogeneous half-space model (*hhs*), middle: layered model with conductive layer (*lay*), right: model with two block anomalies (*bodies*) (Rulff et al., 2021).

Setup:

Source & receiver line aligned in x-direction

receiver $\#$ 1-73	200 to 2000 m
receiver $\#$ 74-146	-200 to -2000 m
dipole source	-50 to 50 m

- Small source and receiver elements
- Reference solution: p2 elements with custEM (Rochlitz et al., 2019)



Example I FE & refinement for a synthetic block model



Element error estimators of the initial mesh (left) and after refinement with error estimator rJ and q = 1.6(right) around the conductive 3D anomaly of the model (Rulff et al., 2021).



Example I FE & refinement for a synthetic block model



Behaviour of error estimators and computational costs (Rulff et al., 2021).



Example II FE & refinement for an ore deposit model



Measurement setup (Rulff et al., 2021)

Model

- 10 000 Ω m homogeneous half space
- Different mineral deposit resistivity scenarios (10, 100, 1 000 Ω m)
- Different values of μ_r for the mineral deposits: 1.3 or 1.0
- Frequency range: 100 Hz 10 kHz
- Source current: 10 A

Purpose: detectability study



Example II FE & refinement for an ore deposit model



Refinement

- Magnetic mineral deposits $(\rho = 1 \ 000 \ \Omega m, \mu_r = 1.3)$
- Error estimator JH
- Volume constraints in the complex-shaped anomalies
- Gradually increasing mesh quality factor



Initial and refined meshes (Rulff et al., 2021).

FE & refinement for metallic infrastructure

Perfect electric conductor - PEC (Alumbaugh and Newman, 1996; Um et al., 2020)

- Metallic-cased well: linewise PEC or combination of solid prism + PEC
- Implemented like a boundary condition on mesh edges corresponding to PEC
- Problem sizes are reduced & arbitrary orientation of wells possible



Applied options for representing a metallic-cased well (Reyes, Rulff et al., talk O2.5.3 @ EMIW 2022)



FE & refinement for metallic infrastructure

Model: Receiver line above basin topography crossing a metallic-cased borehole



Slice through the central model region (Castillo-Reyes et al., 2023, in review)

Model parameters:

- Frequency: 2 Hz
- Receiver spacing: 25 m
- Source: in-line, 560 m distance to well
- One 200 m deep borehole

Modelling codes:

- Refinement: elfe3D (Rulff et al., 2021)
- Higher-order basis functions: PETGEM (Castillo-Reyes et al., 2018)

FE & refinement for metallic infrastructure

Input mesh characteristics

- Fine source and receiver elements
- Well representation: PEC of 25 m long segments
- Mesh quality factor: q= 1.6
- 235 676 elements, 275 747 edges

Refined mesh characteristics

- Refined inner mesh region including PEC
- Mesh quality factor: q= 1.4
- 618 526 elements, 720 565 edges



Element volumes of refined mesh at borehole location (Castillo-Reyes et al., u 2023, in review)



FE & refinement for metallic infrastructure



In-line electric field component (Reyes, Rulff et al., talk O2.5.3 @ EMIW 2022), reference: Castillo-Reyes et al. (2021)



FE & refinement for metallic infrastructure



Performance summary modified after (Castillo-Reyes et al., 2023, in review)

Mesh qualities:

```
Initial I: q = 1.6 (lowest)
Initial II: q = 1.4
Initial III: q = 1.2 (highest)
Refined: q = 1.4
```

PEC representation of casings, goal-oriented refinement and p = 2elements work well together & produce accurate results while minimising problem sizes.

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Spectral elements (SE)



Interpolation functions: *M*-th order curl-conforming basis functions (Duruflé, 2006; Cohen and Duruflé, 2007)

 $\mathbf{N}_{i}^{\xi} = \phi^{(M-1)}(\xi)\phi^{(M)}(\eta)\phi^{(M)}(\zeta)\hat{\xi}$

Code: Spectral element code (Weiss et al., 2023a)

- Variable σ , μ and ϵ distributions
- Arbitrary high-order basis functions based on Lagrange polynomials defined on Gauss and Gauss-Lobatto-Legendre quadrature points
- Solver: MUMPS (Amestoy et al., 2001) or preconditioned iterative algorithm (Weiss et al., 2023b)



Example Buried ore body



Vertical slice at y = 0 (Weiss et al., 2023a)

Model parameters

- Frequency: 100 Hz
- Receiver line of 4 km length with spacing of 100 m
- Source: 400 m long grounded cable perpendicular to receiver line and offset by 500 m
- 2nd order approximation



Example Buried ore body



Comparison of spectral and finite element (custEM; Rochlitz et al., 2019) results (Weiss et al., 2023a)

Example Buried ore body

Comparison of computational quantities between the spectral and a finite element code (custEM; Rochlitz et al., 2019) based on 2nd order basis functions for the buried ore body model (cf. Weiss et al., 2023a)

	Buried ore body model						
	#elements #DOF time						
SEM	90,714	2,200,218	1110.4				
custEM	185,428	2,365,524	775				

- ⇒ Comparable run times for comparable degrees of freedom (DOF)
- ⇒ Fewer elements required for spectral element approach



Topics



- Finite elements with mesh refinement
- Spectral elements with high-order interpolation
- **Boundary Methods**
 - Standard method: Dirichlet
 - Perfectly Matched Layers
- - Direct Solvers
 - Iterative Solvers



Dirichlet BC and source implementation in 2D MT modelling

- BC should ideally absorb/account for all outward directed fields
- Standard: Dirichlet BC
 - ⇒ Prescribed values must exactly match the 2D solution that would be computed at the boundary
- Plane-wave source is represented either at the anomalies or at the outer boundary
- Solution for **1D layered half space** analytically computable





Dirichlet BC for 2D models with different sections at the left and right boundary

Option:

- Prescribe different 1D solutions at the left & right, interpolate horizontally along top
- Field correctly prescribed at sides and bottom, but only approximated at the top
- \longrightarrow Boundary effects from the top affect the solution in the entire domain, also at the receivers





Perfectly Matched Layers



- Absorbing boundary method
- Sequence of artificial layers damping the field to zero in a **reflection-free** manner
- Outer boundary: homogeneous Dirichlet BC
- Damping coefficient: $\eta = \alpha \mid z_{PML} \mid^{\beta}$
- PML may be placed closer to anomalies than Dirichlet boundaries without affecting the solution at the receivers



- PML can directly replace homogeneous Dirichlet BC in the anomalous-field approach (Rivera-Rios et al., 2019; Lei et al., 2022)
- Inhomogeneous BC cannot be directly replaced, instead, we can make use of a method called total and scattered field decomposition (e.g. Lou et al., 2005; Riley et al., 2006; Jin and Riley, 2008; Rumpf, 2012)
- Method originally for antenna studies, we adapted it and developed setups suitable for Earth models



Horizontal TSFD:

- Divide domain along horizontal interface into:
 - Total-field region: contains Earth, solution variable = total field
 - Scattered-field region: contains air layer, solution variable = scattered field
- Incident plane-wave source field excited at the TSFD interface
- PML at top, absorb the scattered field
- \implies useful for realistic models with different sections at the left and right



from: Buntin et al. (under review)



Surrounding TSFD:

- Divide domain along rectangular interface into:
 - Total-field region: contains anomalies, solution variable = total field
 - Scattered-field region: contains horizontally layered background, solution variable = anomalous field
- Normal field (1D background) excited at the interface
- PML at outer boundary, absorb the anomalous field
- \implies useful to shrink the domain for densely discretised high-frequency models







	solution varia		
approach	SFR	TFR	TSFD interface
horizontal TSFD	scattered	total	incident
surrounding TSFD	anomalous	total	normal

Jump at the TSFD interface is implemented by changing the right-hand side vector:

$$r_j = \begin{cases} -A_{jk}u_k & \text{if } j \text{ at TSFD interface, } k \text{ in SFR,} \\ +A_{jk}u_k & \text{if } j \text{ in SFR, } k \text{ at TSFD interface,} \end{cases}$$

(6)



from: Buntin et al. (under review)

 $\forall j, k = 1, 2, 3$ where $u_k = u_k^{\text{incident}}$ (horizontal TSFD) or $u_k = u_k^{\text{normal}}$ (surrounding TSFD).



Topics

Discretisati

- Finite elements with mesh refinement
- Spectral elements with high-order interpolation
- 2 Boundary Methods
 - Standard method: Dirichlet
 - Perfectly Matched Layers

3 Solving

- Direct Solvers
- Iterative Solvers



Solution techniques for systems of equations (SOE)

Linear SOE can be solved using either direct or iterative solvers

Direct solvers

- Very robust and easy to use
- Time and memory complexities of sparse direct solver are O(N²) and O(N^{4/3}) in 3D
- Non-optimal scalability
- ⇒ Use computer clusters to avoid memory bottle neck

Iterative solvers

- Lack robustness
- Only need storage for non-zero entries of system matrix and several additional vectors
- Rely on easily scalable matrix-vector multiplications and vector operations
- ⇒ Use preconditioning to ensure robustness and improve convergence rate



Prerequisites for iterative solution

Rewrite complex-valued SOE

 $\left(\textbf{K}+\textit{i}\textbf{M}_{\sigma}-\textbf{M}_{\epsilon}\right)\textbf{E}=\textbf{b},$

as real-equivalent two-by-two system of form

$$\underbrace{\begin{bmatrix} \textbf{M}_{\sigma} & -(\textbf{K}-\textbf{M}_{\epsilon}) \\ \textbf{K}-\textbf{M}_{\epsilon} & \textbf{M}_{\sigma} \end{bmatrix}}_{\textbf{C}_{\text{RI}}} \begin{bmatrix} \textbf{E}_{\textbf{R}} \\ -\textbf{E}_{\textbf{I}} \end{bmatrix} = \begin{bmatrix} \textbf{b}_{\textbf{I}} \\ \textbf{b}_{\textbf{R}} \end{bmatrix},$$



PREconditioning for Square Blocks: PRESB

PRESB is a robust and efficient preconditioner for two-by-two block systems of general form

$$\mathcal{A} = egin{bmatrix} \mathbf{A} & -b \, \mathbf{B}_2 \ a \mathbf{B}_1 & \mathbf{A} \end{bmatrix},$$

with the assumption that A is symmetric positive definite. PRESB reads as follows

$$\mathscr{P} = egin{bmatrix} \mathsf{A} & -\mathsf{B}_2 \ \mathsf{B}_1 & \mathsf{A} + \sqrt{ab}(\mathsf{B}_1 + \mathsf{B}_2) \end{bmatrix}$$

and it can be proved that all the eigenvalues of the preconditioned system $\mathcal{P}^{-1}\mathcal{A}$ are clustered in the interval $[\frac{1}{2}, 1]$, **independently of the discretisation and material parameters** and the scalars *a* and *b* (see Axelsson et al., 2016).



To solve

$$\begin{bmatrix} \mathbf{M}_{\sigma} & -(\mathbf{K} - \mathbf{M}_{\epsilon}) \\ \mathbf{K} - \mathbf{M}_{\epsilon} & \mathbf{M}_{\sigma} \end{bmatrix} \begin{bmatrix} \mathbf{E}_{\mathbf{R}} \\ -\mathbf{E}_{\mathbf{I}} \end{bmatrix} = \begin{bmatrix} \mathbf{b}_{\mathbf{I}} \\ \mathbf{b}_{\mathbf{R}} \end{bmatrix}$$

the Generalized Conjugate Residual (GCR) method is applied. To precondition this system, PRESB is taken as preconditioner

$$\mathbf{P} = \begin{bmatrix} \mathbf{M}_{\sigma} & -(\mathbf{K} - \mathbf{M}_{\epsilon}) \\ \mathbf{K} - \mathbf{M}_{\epsilon} & \mathbf{M}_{\sigma} + 2(\mathbf{K} - \mathbf{M}_{\epsilon}) \end{bmatrix}.$$



Preconditioner PRESB

The computations of applying the PRESB preconditioner, that is solving a system of form

$$\mathsf{P}\begin{bmatrix}\mathsf{w}_1\\\mathsf{w}_2\end{bmatrix} = \begin{bmatrix}\mathsf{f}_1\\\mathsf{f}_2\end{bmatrix}$$

consists of performing the following steps:

Algorithm 1: Solving linear system with preconditioner P

- 1 Set $\mathbf{H} = \mathbf{M}_{\sigma} + (\mathbf{K} \mathbf{M}_{\epsilon})$
- ${\scriptstyle 2}\;$ Solve $Hg=f_1+f_2$
- $_{3}$ Compute $M_{\sigma}g$ and $f_{1}-M_{\sigma}g$
- 4 Solve $Hh = f_1 M_\sigma g$
- s Compute $w_1 = g + h$ and $w_2 = -h$



Iterative algorithm

Algorithm 2: Iterative framework

Input: C_{RI} , b, H, M_{σ} , tol

Output: E

- 1 Let \mathbf{x}_0 be the initial guess
- ${}_{\mathbf{2}} \;\; \operatorname{Set} {\boldsymbol{r}}_0 = {\boldsymbol{b}} {\boldsymbol{C}}_{\boldsymbol{\mathsf{RI}}} {\boldsymbol{x}}_0$
- ${}_3$ for $i=0,\ldots m$ do
- 4 Compute \mathbf{p}_i using Algorithm 1 with \mathbf{r}_i being the right hand side

$$\begin{array}{c|c} \mathbf{s} & \mathbf{q}_{i} \leftarrow \mathbf{C}_{\mathbf{R}\mathbf{I}}\mathbf{p}_{i} \\ \mathbf{6} & \mathbf{q}_{i} \leftarrow \mathbf{q}_{i} - \sum_{j=0}^{i-1} \mathbf{q}_{j} \frac{(\mathbf{q}_{i}, \mathbf{q}_{i})}{(\mathbf{q}_{i}, \mathbf{q}_{i})} \\ \mathbf{7} & \mathbf{p}_{i} \leftarrow \mathbf{p}_{i} - \sum_{j=0}^{i-1} \mathbf{p}_{j} \frac{(\mathbf{q}_{i}, \mathbf{q}_{i})}{(\mathbf{q}_{i}, \mathbf{q}_{i})} \\ \mathbf{8} & al_{i} \leftarrow \frac{(\mathbf{r}_{i}, \mathbf{q}_{i})}{(\mathbf{q}_{i}, \mathbf{q}_{i})} \\ \mathbf{9} & \mathbf{x}_{i+1} \leftarrow \mathbf{x}_{i} + al_{i}\mathbf{p}_{i} \\ \mathbf{10} & \mathbf{r}_{i+1} \leftarrow \mathbf{r}_{i} - al_{i}\mathbf{q}_{i} \end{array}$$

11 **if**
$$\frac{||\mathbf{r}_{i+1}||_2}{||\mathbf{b}||_2} < tol$$
 then Stop

12 | 13 end



Implementation

- Code written in Fortran
- Implemented in a parallel fashion using Message Passing Interface (MPI)
- Relying on functionalities of open-source libraries PETSc (Balay et al., 1997, 2022), MUMPS (Amestoy et al., 2001, 2006) and hypre (Falgout and Yang, 2002; Falgout et al., 2006).
- Outer-inner solution method
- Outer solver: GCR preconditioned with PRESB
- Inner solver: GCR or flexible generalised minimal residual (FGMRES) method preconditioned with the auxiliary-space Maxwell solver (AMS; Kolev and Vassilevski, 2006, 2009; Grayver and Kolev, 2015)



Test models



Schematic side views (Weiss et al., 2023b)



Test models

Model information (see Weiss et al., 2023b)

Model	Layered Earth	3-D model
Domain size [km ³]	$30\times 30\times 30$	$30\times 36\times 30$
Conductivities [S/m]	$\label{eq:shift} \begin{split} \sigma_{Air} &= 10^{-8}, \sigma_{Earth} = 10^{-4}, \\ \sigma_{Layer} &= 10^{-2} \end{split}$	$\label{eq:starth} \begin{split} \sigma_{Air} &= 10^{-8}, \sigma_{Earth} = 10^{-4}, \\ \sigma_{cover} &= 0.01, \sigma_{ore} = 1 \end{split}$
Approximation order	1st	1st
# elements	$54\times54\times54$	332'580
# degrees of freedom	980'100	2'033'986

- Runs performed on AMD Ryzen Threadripper 2950X 16-core processor clocked at 3.5 GHz and with 128 GB RAM
- Run with two MPI processes



Robustness of solver with respect to frequency



Convergence of outer solver (Weiss et al., 2023b)



Robustness of solver with respect to frequency

Computational quantities for the **layered Earth model** and variable frequencies for edge elements of degree p = 1. The number of outer GCR iterations is denoted by N_{it}^{outer} (Weiss et al., 2023b)

Inner solver	AMS-GCR			AMS-FGMRES			DMUMPS		
freq [Hz]	N ^{outer}	time [s]	mem [GB]	N ^{outer}	time [s]	mem [GB]	N ^{outer}	time [s]	mem [GB]
0.1	7	40.0	4.3	7	39.3	4.3	6	121.8	14.1
1	10	56.7	4.4	10	55.8	4.3	9	147.3	14.1
10	16	79.3	4.4	16	78.3	4.4	15	145.3	14.0
100	22	93.5	4.5	22	92.7	4.3	18	157.2	14.1
1000	19	70.2	4.4	19	69.3	4.3	17	166.7	14.0
5000	18	62.8	4.4	18	62.1	4.3	17	153.0	14.0
8000	18	96.8	4.4	18	95.6	4.4	18	157.8	14.2
10000	18	460.7	4.4	18	443.4	4.4	18	168.7	14.1 UPPS

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Robustness of solver with respect to frequency



Convergence curves of the inner AMS-preconditioned GCR solver for the first inner system (Steps 1 of Algorithm 1) when the normalised relative residual of the outer solver reaches values of 10^{-4} and 10^{-8} (Weiss et al., 2023b)

Robustness of solver with respect to problem size

Comparison of outer iteration counts (N^{outer}) and solving times (time [s]) for various frequencies and problem sizes **for the layered Earth model** (Weiss et al., 2023b)

	frequency [Hz]							
	(0.1 10		1000		8000		
#DOF	N ^{outer}	time[s]						
980'100	7	42.2	16	79.3	19	70.2	18	96.8
3'641'400	8	152.9	15	286.4	18	272.6	19	310.2
6'879'600	8	343.4	16	646.5	18	521.8	18	790.5

- Stable outer iteration count with respect to problem size
- Time increases about linearly



Robustness of solver with respect to permeability

Comparison of outer iteration counts (N^{outer}) and solving times (time [s]) for various frequencies and different magnetic permeability values for the ore body of the **3D** model (Weiss et al., 2023b)

	relative magnetic permeability μ_r of ore body							
	1		2		5		10	
frequency [Hz]	N ^{outer}	time[s]	N ^{outer}	time[s]	N ^{outer}	time[s]	N ^{outer}	time[s]
0.1	9	594.5	9	586.0	10	656.7	11	691.8
10	16	284.8	16	282.3	16	180.1	16	279.2
100	23	278.6	23	275.4	23	273.7	24	284.1
8000	18	324.3	18	345.0	18	344.7	18	331.4

Robustness of solver with respect to permeability & permittivity

Comparison of iteration counts (N_{it}^{outer}) and solving times (time [s]) for various frequencies. Note that these simulations are run with a variable dielectric permittivity distribution and two different magnetic permeability values for the ore body of the **3D model** (Weiss et al., 2023b)

relative dielectric permittivity	eair -	Ecover -	20 Ehost r	$ock = 5 e^{ore body} = 1$	
of air, cover, host rock and ore body	$c_r =$, c _r –	$L0, c_r$	$=$ 0, e_r $=$ 1	
relative magnetic permeability		<u> </u>		<i>u</i> – 10	
of ore body	$\mu_r = 1$		$\mu_r = 10$		
frequency [Hz]	N ^{outer}	time[s]	N ^{outer}	time[s]	
0.1	9	590.4	11	691.6	
10	16	282.3	16	279.8	
100	23	278.5	24	285.1	
8000	18	341.0	18	322.4	

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Computational comparisons of iterative and direct solvers in terms of time and memory consumption for the **layered Earth model** at a frequency of 100 Hz for different problem sizes (Weiss et al., 2023b)

		Iterative Me	Direct S	olver: ZMUMPS		
Inner solver	Preconditioned GCR		DMUMPS			-
#DOF	time[s]	mem[GB]	time[s]	mem[GB]	time[s]	mem[GB]
980'100	93.5	4.4	157.2	14.1	166.8	9.0
3'641'400	368.4	16.4	1625.0	74.8	1874.1	55.5
6'879'600	661.0	30.1	-	out of memory	-	out of memory

- Iterative solver requires far less memory than direct solver
- Iterative framework also needs less time to solve linear system of equations



Take aways

Technical insights into advanced numerical methods applied in 3D EM modelling

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Efficient goal-oriented mesh refinement in 3-D finite-element modelling adapted for controlled source electromagnetic surveys

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Spectral element method for 3-D controlled-source electromagnetic forward modelling using unstructured hexahedral meshes

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Improved accuracy of plane-wave electromagnetic modelling by application of the total and scattered field decomposition and perfectly matched layers

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ORIGINAL PAPER



Iterative solution methods electromagnetic forward modelling of geophysical exploration scenarios

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Contributions

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- MT FE modelling
- Boundary methods
- TSFD

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- CSEM SE modelling
- Iterative solution methods

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- CSEM FE modelling
- Mesh refinement
- Inversion



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